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- Review: arbitrary pole placement by full state feedback.
- Today's topic: observer design for state estimation when full state feedback is not implementable.


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Reading: FPE, Chapter 7

## Review: Pole Placement via State Feedback

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Closed-loop poles are the eigenvalues of $A-B K!!$

Review: Pole Placement in CCF

$$
\begin{aligned}
& \dot{x}=(A-B K) x+B r, \quad y=C x \\
& A-B K=-\left(\begin{array}{ccccc}
0 & 1 & \ldots & 0 & 0 \\
0 & 0 & \ldots & 0 & 0 \\
\vdots & \vdots & \ddots & \vdots & \vdots \\
0 & 0 & \ldots & 0 & 1 \\
a_{n}+k_{1} & a_{n-1}+k_{2} & \ldots & a_{2}+k_{n-1} & a_{1}+k_{n}
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\end{aligned}
$$

Closed-loop poles are the roots of the characteristic polynomial

$$
\begin{aligned}
& \operatorname{det}(I s-A+B K) \\
& =s^{n}+\left(a_{1}+k_{n}\right) s^{n-1}+\ldots+\left(a_{n-1}+k_{2}\right) s+\left(a_{n}+k_{1}\right)
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Key observation: When the system is in CCF, each control gain affects only one of the coefficients of the characteristic polynomial, and these coefficients can be assigned arbitrarily by a suitable choice of $k_{1}, \ldots, k_{n}$.

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Key observation: When the system is in CCF, each control gain affects only one of the coefficients of the characteristic polynomial, and these coefficients can be assigned arbitrarily by a suitable choice of $k_{1}, \ldots, k_{n}$.

Hence the name Controller Canonical Form - convenient for control design.

## Pole Placement by State Feedback

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3. Convert back to original coordinates.

## Is Full State Feedback Always Available?

In a typical system, measurements are provided by sensors:


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Full state feedback $u=-K x$ is not implementable!!

## When Full State Feedback Is Unavailable ...

... we need an Observer!!

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## State Estimation Using an Observer

When full state feedback is unavailable, the observer is used to estimate the state $x$ :


## State Estimation Using an Observer

The idea is to design the observer in such a way that the state estimate $\widehat{x}$ is asymptotically accurate:

$$
\|\widehat{x}(t)-x(t)\|=\sqrt{\sum_{i=1}^{n}\left(\widehat{x}_{i}(t)-x_{i}(t)\right)^{2}} \xrightarrow{t \rightarrow \infty} 0
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If we are successful, then we can try estimated state feedback:


A New Concept: Observability

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- Before, we saw that closed-loop poles can be assigned arbitrarily by full state feedback when the plant is controllable.
- Now, we will see that asymptotically accurate state estimation will be possible when the system is observable.
- Observability is a system property which is dual to controllability.


## Observability

Consider a single-output system $(y \in \mathbb{R})$ :

$$
\dot{x}=A x+B u, \quad y=C x \quad x \in \mathbb{R}^{n}
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$$
\mathcal{O}(A, C)=\left[\begin{array}{c}
C \\
C A \\
\vdots \\
C A^{n-1}
\end{array}\right]
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- recall that $C$ is $1 \times n$ and $A$ is $n \times n$, so $\mathcal{O}(A, C)$ is $n \times n$;


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We say that the above system is observable if its observability matrix $\mathcal{O}(A, C)$ is invertible.

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We say that the above system is observable if its observability matrix $\mathcal{O}(A, C)$ is invertible.
(This definition is only true for the single-output case; the multiple-output case involves the rank of $\mathcal{O}(A, C)$.)

## Example: Computing $\mathcal{O}(A, C)$

$$
\text { Let } \quad A=\left(\begin{array}{ll}
0 & -6 \\
1 & -5
\end{array}\right), \quad C=\left(\begin{array}{ll}
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Here, $n=2, C \in \mathbb{R}^{1 \times 2}, A \in \mathbb{R}^{2 \times 2} \Longrightarrow \mathcal{O}(A, C) \in \mathbb{R}^{2 \times 2}$.

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\end{array}\right)=\left(\begin{array}{ll}
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\therefore \mathcal{O}(A, C)=\left(\begin{array}{cc}
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\end{gathered}
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$\therefore \mathcal{O}(A, C)=\left(\begin{array}{cc}0 & 1 \\ 1 & -5\end{array}\right)$ $\operatorname{det} \mathcal{O}(A, C)=-1$ $\Longrightarrow$

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$\therefore \mathcal{O}(A, C)=\left(\begin{array}{cc}0 & 1 \\ 1 & -5\end{array}\right)$
$\operatorname{det} \mathcal{O}(A, C)=-1 \quad \Longrightarrow \quad$ the system is observable

- recall: this system is in Observer Canonical Form (OCF) ...


## Observer Canonical Form

A single-output state-space model

$$
\dot{x}=A x+B u, \quad y=C x
$$

is said to be in Observer Canonical Form (OCF) if the matrices $A, C$ are of the form

$$
A=\left(\begin{array}{cccccc}
0 & 0 & \ldots & 0 & 0 & * \\
1 & 0 & \ldots & 0 & 0 & * \\
\vdots & \vdots & \ddots & \vdots & \vdots & \vdots \\
0 & 0 & \ldots & 1 & 0 & * \\
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Fact: A system in OCF is always observable!!

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\end{array}\right), \quad C=\left(\begin{array}{lllll}
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Fact: A system in OCF is always observable!!
(The proof of this for $n>2$ uses the Jordan canonical form, we will not worry about this.)

## Coordinate Transformations and Observability

Just like controllability, observability is preserved under invertible coordinate transformations.

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$$
\begin{array}{lrr}
\dot{x} & =A x+B u & \xrightarrow{T} \\
y & =C x & \dot{\bar{x}}=\bar{A} \bar{x}+\bar{B} u \\
y & =\bar{C} \bar{x}
\end{array}
$$

where $\bar{A}=T A T^{-1}, \quad \bar{B}=T B, \quad \bar{C}=C T^{-1}$

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y & =C x & y
\end{array}
$$

where $\bar{A}=T A T^{-1}, \quad \bar{B}=T B, \quad \bar{C}=C T^{-1}$

$$
\mathcal{O}(\bar{A}, \bar{C})=\left(\begin{array}{c}
\bar{C} \\
\bar{C} \bar{A} \\
\vdots \\
\bar{C} \bar{A}^{n-1}
\end{array}\right)
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y & =C x
\end{array} \quad y=\bar{C} \bar{x} \text { and }
$$

where $\bar{A}=T A T^{-1}, \quad \bar{B}=T B, \quad \bar{C}=C T^{-1}$

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\end{array}\right)=\left(\begin{array}{c}
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\vdots \\
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C T^{-1} T A T^{-1} \\
\vdots \\
C T^{-1} T A^{n-1} T^{-1}
\end{array}\right) \\
& =\left(\begin{array}{c}
C \\
C A \\
\vdots \\
C A^{n-1}
\end{array}\right) T^{-1}=\mathcal{O}(A, C) T^{-1}
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where $\bar{A}=T A T^{-1}, \quad \bar{B}=T B, \quad \bar{C}=C T^{-1}$
If the original system is observable, then

$$
\begin{aligned}
& T \underbrace{[\mathcal{O}(A, C)]^{-1}}_{\text {old }}=\underbrace{[\mathcal{O}(\bar{A}, \bar{C})]^{-1}}_{\text {new }} \\
& T=\underbrace{[\mathcal{O}(\bar{A}, \bar{C})]^{-1}}_{\text {new }} \underbrace{[\mathcal{O}(A, C)]}_{\text {old }}
\end{aligned}
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## Observability and State Estimation

As we will show next:

If the system is observable, then there exists an observer (state estimator) that provides an asymptotically convergent estimate $\widehat{x}$ of the state $x$ based on the observed output $y$.

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The particular type of observer we will construct is called the Luenberger observer after David G. Luenberger, who developed this idea in his 1963 Ph.D. dissertation.

David Luenberger is a Professor at Stanford University.

## The Luenberger Observer

Consider a state-space model

$$
\begin{aligned}
& \dot{x}=A x \quad(\text { for now, assume } u=0) \\
& y=C x
\end{aligned}
$$

## The Luenberger Observer

Consider a state-space model

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$$

We wish to estimate the state $x$ based on the output $y$.
Consider feeding the output $y$ as input to the following system with state $\widehat{x}$ :

$$
\dot{\hat{x}}=(A-L C) \widehat{x}+L y
$$

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Consider feeding the output $y$ as input to the following system with state $\widehat{x}$ :

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\dot{\hat{x}}=(A-L C) \widehat{x}+L y
$$

Assumption: The output injection matrix $L$ is chosen in such a way that the matrix $A-L C$ is Hurwitz (i.e., all of its eigenvalues lie in LHP).

## The Luenberger Observer

Consider a state-space model

$$
\begin{aligned}
& \dot{x}=A x \quad(\text { for now, assume } u=0) \\
& y=C x
\end{aligned}
$$

We wish to estimate the state $x$ based on the output $y$.
Consider feeding the output $y$ as input to the following system with state $\widehat{x}$ :

$$
\dot{\hat{x}}=(A-L C) \widehat{x}+L y
$$

Assumption: The output injection matrix $L$ is chosen in such a way that the matrix $A-L C$ is Hurwitz (i.e., all of its eigenvalues lie in LHP).

At this point, we do not assume anything about observability.

## The Luenberger Observer

System: $\quad \dot{x}=A x$

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y=C x
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\dot{e}=\dot{x}-\dot{\widehat{x}}
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What happens to state estimation error $e=x-\widehat{x}$ as $t \rightarrow \infty$ ?

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& =A x-[(A-L C) \widehat{x}+L C x]
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Does $e(t)$ converge to zero in some sense?

## Linear ODEs and Eigenvalues: A Digression

$$
\dot{v}=F v, \quad v \in \mathbb{R}^{n}, F \in \mathbb{R}^{n \times n}
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Then there exists a matrix $T \in \mathbb{R}^{n \times n}$, such that $T^{-1}=T^{T}$ and

$$
F=T^{-1}\left(\begin{array}{llll}
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& \lambda_{2} & & \\
& & \ddots & \\
& & & \lambda_{n}
\end{array}\right) T
$$

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$$

$$
\Uparrow
$$

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\dot{\bar{v}}_{i}=\lambda_{i} \bar{v}_{i}, \quad i=1,2, \ldots, n
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This system of $n$ 1st-order ODEs has the solution

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If all $\lambda_{i}$ 's have negative real parts, then

$$
\begin{aligned}
\|v(t)\|^{2} & =v(t)^{T} v(t)=\bar{v}(t)^{T} \bar{v}(t) \\
& \leq C e^{-2 \sigma_{\min } t}, \quad \text { where } \sigma_{\min }=\min _{1 \leq i \leq n}\left|\operatorname{Re}\left(\lambda_{i}\right)\right|
\end{aligned}
$$

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Observer:

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& \dot{\widehat{x}}=(A-L C) \widehat{x}+L y \\
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Recall our assumption that $A-L C$ is Hurwitz (all eigenvalues are in LHP). This implies that

$$
\|x(t)-\widehat{x}(t)\|^{2}=\|e(t)\|^{2}=\sum_{i=1}^{n}\left|e_{i}(t)\right|^{2} \xrightarrow{t \rightarrow \infty} 0
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at an exponential rate, determined by the eigenvalues of $A-L C$.

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For fast convergence, want eigenvalues of $A-L C$ far into LHP!!

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Error:
Observer transfer function:

$$
\begin{aligned}
& s \widehat{X}(s)=(A-L C) \widehat{X}(s)+L Y(s) \\
& (I s-A+L C) \widehat{X}(s)=L Y(s) \\
& \widehat{X}(s)=(I s-A+L C)^{-1} L Y(s)
\end{aligned}
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$$

The eigenvalues of $A-L C$ are the observer poles. We want these poles to be stable and fast.

## Observability and Estimation Error

Fact: If the system

$$
\dot{x}=A x, \quad y=C x
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is observable, then we can arbitrarily assign eigenvalues of $A-L C$ by a suitable choice of the output injection matrix $L$.

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This is similar to the fact that controllability implies arbitrary closed-loop pole placement by state feedback.

In fact, these two facts are closely related because CCF is dual to OCF.

## Observer Pole Placement in OCF

Consider a single-output system in OCF:

$$
\begin{aligned}
& \dot{x}=A x \\
& y=C x, \quad y \in \mathbb{R}
\end{aligned}
$$

where $A=\left(\begin{array}{cccccc}0 & 0 & \ldots & 0 & 0 & -a_{n} \\ 1 & 0 & \ldots & 0 & 0 & -a_{n-1} \\ \vdots & \vdots & \ddots & \vdots & \vdots & \vdots \\ 0 & 0 & \ldots & 1 & 0 & -a_{2} \\ 0 & 0 & \ldots & 0 & 1 & -a_{1}\end{array}\right), \quad C=\left(\begin{array}{lllll}0 & 0 & \ldots & 0 & 1\end{array}\right)$

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Note that $A^{T}$ has the form of a CCF system matrix, thus:

$$
\begin{aligned}
\operatorname{det}(I s-A) & =\operatorname{det}\left((I s-A)^{T}\right)=\operatorname{det}\left(I s-A^{T}\right) \\
& =s^{n}+a_{1} s^{n-1}+\ldots+a_{n-1} s+a_{n}
\end{aligned}
$$

## Now Let's Add an Observer

$$
A=\left(\begin{array}{ccccc}
0 & 0 & \ldots & 0 & -a_{n} \\
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L C & =\left(\begin{array}{c}
\ell_{1} \\
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\vdots \\
\ell_{n}
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& A-L C=\left(\begin{array}{ccccc}
0 & 0 & \ldots & 0 & -\left(a_{n}+\ell_{1}\right) \\
0 & 1 & \ldots & 0 & -\left(a_{n-1}+\ell_{2}\right) \\
\vdots & \vdots & \ddots & \vdots & \vdots \\
0 & 0 & \ldots & 0 & -\left(a_{2}+\ell_{n-1}\right) \\
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Eigenvalues of $A-L C$ are the roots of the characteristic polynomial

$$
\begin{aligned}
& \operatorname{det}(I s-A+L C) \\
& =s^{n}+\left(a_{1}+\ell_{n}\right) s^{n-1}+\ldots+\left(a_{n-1}+\ell_{2}\right) s+\left(a_{n}+\ell_{1}\right)
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Key observation: In OCF, each observer gain affects only one of the coefficients of the characteristic polynomial, which can be assigned arbitrarily by a suitable choice of $\ell_{1}, \ldots, \ell_{n}$.

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Hence the name Observer Canonical Form - convenient for observer design.

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General procedure for any observable system:

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The resulting observer is

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In fact, this procedure is not necessary because of duality between controllability and observability!!

## Controllability-Observability Duality

Claim: The system

$$
\dot{x}=A x, \quad y=C x
$$

is observable if and only if the system

$$
\dot{x}=A^{T} x+C^{T} u
$$

is controllable.

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Thus, $\mathcal{O}(A, C)$ is nonsingular if and only if $\mathcal{C}\left(A^{T}, C^{T}\right)$ is.

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Final answer: use the observer

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\begin{aligned}
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& =\left(A-K^{T} C\right) \widehat{x}+K^{T} y .
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